

# A Methodology for Obtaining Voltage and Current Ripples of Power Electronics Converters with a Fixed Node on their Output

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## Abstract

This paper presents a methodology for obtaining voltage and current ripples for power electronics (PE) converters which have a fixed node on its output. The operation principle of PE converters basically consists of closing and opening a power switch for regulating the voltages and currents of the system according to their control objectives. In this paper, the analysis is performed with PE converters with a node in the output which topology does not change for both switching states; in other words, the differential equations that govern the behavior of output current do not depend on the duty cycle. According to the literature review, most researches focus on modelling and controlling PE converters; however, to the best of the authors' knowledge, a detailed methodology for deducing ripples of PE converters has not yet been reported. Ripples are necessary for appropriately sizing the passive elements in the design stage. The methodology is explained using the Buck converter; nonetheless, it can be applied for any PE converter that has a fixed node on its output. Finally, OpenModelica software is used for validating the proposed methodology.

**Keywords:** Voltage and current ripples, Power Electronics (PE) devices, Differential Equations (DE), OpenModelica.

## I. INTRODUCTION

Deduction of voltage and current ripples are useful in the design stage of a converter for correctly sizing passive elements, basically inductors ( $L$ ) and capacitors ( $C$ ). Engineers, in the design stage, must comply with ripple requirements that loads or applications may have. Voltage and current ripples depend on: values of inductor and capacitors, switching frequency, duty cycle and output voltage. Usually, current ripples equations are easy to determine; however, the deduction of output voltage ripples equations has a greater degree of difficulty for two reasons: 1) output capacitor stores and delivers energy during the same switching state, for both closed and open switching states. 2) Voltage ripple depends on current ripple. This paper presents a methodology for obtaining voltage and current ripples in converters that have a fixed node on their output.

Using the geometric representation of voltage and current ripples, this paper develops a rigorous explanation for deducing the mathematical expression that governs the behavior of ripples in converters. The explanation is presented for giving specific details that can be useful for electrical, control and electronic engineers. The contributions of most power electronic design papers are related to the use of the model, failing to present details regarding the deduction of aspects such as voltage and current ripples. However, such details are of paramount importance, especially for PE designers, since references allow establishing the design requirements in the designing stage [1]-[8].

The dynamical performance can be understood if the operation principle and ripples are explained in detail, which may lead to better setting the requirements established from the design stage [9]. There are a lot of papers that partially include the explanation of the obtention of voltage and current ripples; nonetheless, there are still many gaps in the knowledge [10]-[12]. The main contribution of this paper lies on the deduction of voltage and current ripples that permits a deeper explanation of the operation principle for PE devices. The proposed methodology can be easily applied to other type of converters [13]-[14]. After the deduction of voltage and current ripples, the proposed procedure is validated using the OpenModelica software.

OpenModelica [15] is open source software designed for the simulation of PE devices, allowing dynamic multi-domain simulation of linear and non-linear systems. OpenModelica is made of an equation-based and object-oriented language known as *Modelica*. OpenModelica features extensive model libraries in several fields, as well as other resources such as a graphic connection editor (OMEdit), a compiler, a simulator and plotting tools. OpenModelica is widely used in industrial and research applications on electric and electronic engineering. The authors in [16] used the *Modelica* language in studies of DC microgrids. In [17] the authors simulated electrical power networks, proposing a Power System Library; also, in [18] control algorithms are implemented using OpenModelica for power inverter applications. The aforementioned studies evidence that OpenModelica is a tool for improving the formation of future engineers [19]. In this paper, OpenModelica is used as a validation tool for verifying the obtained reference voltages and currents.

This paper is organized as follows: Section II presents the proposed methodology using the Buck converter. Section III corresponds to the results obtained in OpenModelica for validating the proposed methodology. Section IV concludes and highlights the most relevant aspect of this paper.

## II. PROPOSED METHODOLOGY

The proposed methodology is explained using the Buck converter depicted in Figure 1.  $v_i$  is the input voltage;  $L$  is an inductor;  $C$  is a capacitor;  $Q$  is a power switch,  $D$  is a power diode; and  $R$  is the resistors used as a load. The following assumptions have been made: 1) Losses are neglected. 2) voltage drop in power switches is not considered. 3) The passive law of sign is used to obtain voltage and current references of the system that are drawn with red color.

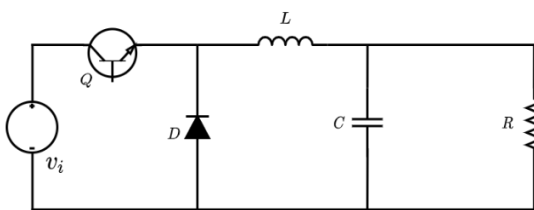


Figure 1. Buck converter topology

Figure 2 shows the converter for both switching states. Figure 2(a) corresponds to  $Q$  closed while Figure 2(b) corresponds to  $Q$  open. Note that for both switching states, the node on the output does not change its topology (fixed node).

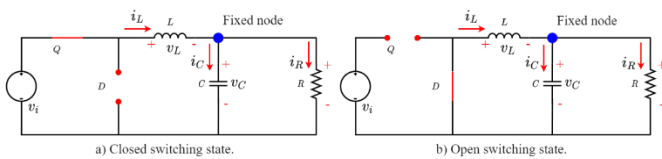


Figure 2. Switching states of the buck converter

Equations (1) and (2) were obtained after applying Kirchoff laws in Figure 2(a) while Equations (3) and (4) were obtained after applying Kirchoff laws in Figure 2(b).

$$L \frac{di_L}{dt} = v_i - v_c \quad (1)$$

$$C \frac{di_C}{dt} = i_L - i_R = i_L - \frac{v_c}{R} \quad (2)$$

$$L \frac{di_L}{dt} = -v_c \quad (3)$$

$$C \frac{di_C}{dt} = i_L - i_R = i_L - \frac{v_c}{R} \quad (4)$$

Please note that Equation (2) and (4) are the same. This is due to the fixed node on the output. Figure 3 correspond to the

voltage and current waveforms of the converter. Figure 3(a) is the Pulse Width Modulation (PWM) signal for controlling the converter.  $Q$  is closed between 0 and  $DT_{sw}$  while is open in the complementary period  $((1-D)T_{sw})$  between  $DT_{sw}$  and  $T_{sw}$ . Figure 3(b) is the waveform of  $i_L$ , being  $I_L$  its mean value and  $\Delta i_L$  its ripple. Figure 3(c) is the current waveform of the load which is assumed constant. Figure 3(d) is the capacitor current ( $i_C$ ) and  $\Delta i_C$  is its respective ripple, for a correct operation of the converter its mean value must be zero ( $I_C = 0$ ). Figure 3(e) is the capacitor voltage waveform ( $v_C$ ),  $\Delta v_C$  is its ripple and  $V_C$  is its mean value.

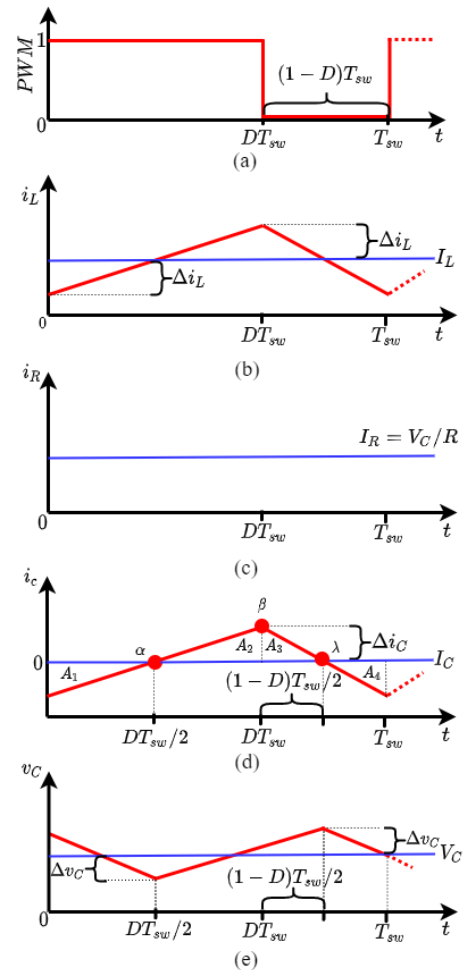


Figure 3. Voltages, current, and ripples representation.

The current ripple ( $\Delta i_L$ ) can be obtained from Equations (1) and (3); however, Equation (3) is simpler. Then, Equation (3) is discretized ( $L \frac{\Delta i_L}{\Delta t} = -V_C$ ) with  $\Delta t = (1-D)T_{sw}$  where  $D$  is the duty cycle and  $T_{sw}$  is the switching period. The current ripple must be divided by 2 since ripples are usually obtained with respect to the mean value of the signal. Equation (5) indicates the current ripple, where  $V_C$  and  $I_L$  are the mean values of  $v_C$  and  $i_L$ , respectively; while  $f_{sw}$  is the switching frequency.

$$\Delta i_L = \frac{V_C(1-D)T_{sw}}{2L} = \frac{V_C(1-D)}{2Lf_{sw}} \quad (5)$$

The correct operation of the converter establishes that the mean value of  $I_C$  must be zero. In consequence, areas  $A_1$  and  $A_2$  in Figure 3(d) must be equal;  $A_3$  and  $A_4$  are also equal. Please note, that zero crossing when  $Q$  is closed is in the time instant of  $(DT_{sw}/2)$  ( $\alpha$  point) while zero crossing when  $Q$  is open is in the time instant of  $(DT_{sw} + (1 - D)T_{sw}/2)$  ( $\lambda$  point). The deduction of the voltage ripple in the capacitor ( $\Delta v_C$ ) considers the following assumptions: 1) the load current ( $I_R$ ) is constant. 2) the inductor current ripple is equal to the capacitor current ripple ( $\Delta i_L = \Delta i_C$ ). 3) voltages and currents increase and decrease linearly.  $\Delta v_C$  can be deduced using the triangle  $\alpha\beta\lambda$  ( $A_{\alpha\beta\lambda}$ ) of Figure 3(d):

$$\int i_C dt = A_2 + A_3 = A_{\alpha\beta\lambda} = \frac{1}{2} \cdot \text{base} \cdot \text{height} \quad (6)$$

$$= \frac{1}{2} \cdot \text{base} \cdot \text{height}$$

Where *base* and *height* are respectively the base and the height of triangle  $A_{\alpha\beta\lambda}$ . The base of the triangle (segment  $\alpha\lambda$ ) is  $(\lambda - \alpha = [DT_{sw} + (1 - D)T_{sw}/2] - [DT_{sw}/2]) = T_{sw}/2$  while its height is  $\Delta i_C$ .

$$\int C \frac{dv_C}{dt} dt = \int_{V_C - \Delta v_C}^{V_C + \Delta v_C} dv_C = \frac{\Delta i_C}{4f_{sw}} \quad (7)$$

Replacing (5) in (7) the expression given in Equation (8) it is obtained:

$$\Delta v_C = \frac{V_C(1 - D)}{16f_{sw}^2 LC} \quad (8)$$

#### IV. RESULTS

A simulation was carried out using the default compiler and solver in OpenModelica connection editor (OMEdit), version 3.2.2. Figure 4 show the implementation in OpenModelica of the converter depicted in Figure 1. The duty cycle used is 50% while the switching frequency is 1 kHz. The Buck converter was parametrized as follows:  $L = 10 \text{ mH}$ ,  $C = 1000 \text{ } \mu\text{F}$ ,  $R = 10 \text{ } \Omega$ ,  $v_i = 100 \text{ V}$ .

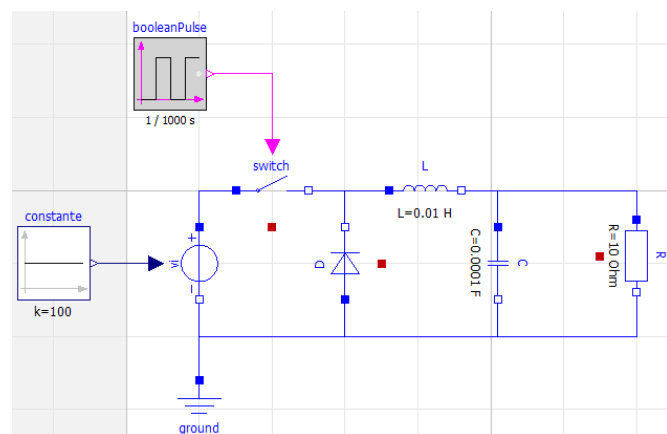


Figure 4. Implementation in OpenModelica.

Figures 5 and 6 show the voltage and current waveforms, respectively, with their respective ripples obtained with the simulation. Note that the theoretical calculation and simulation match, which validates the proposed methodology.

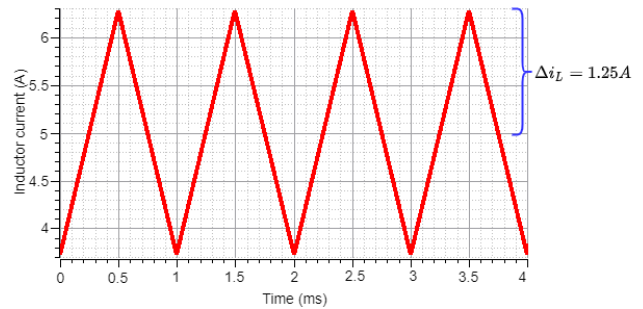


Figure 5. Current waveform and its ripple.

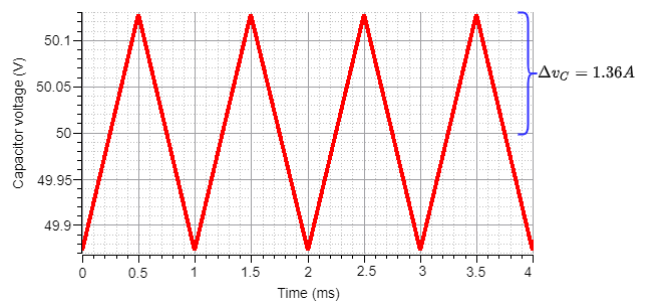


Figure 6. Voltage waveform and its ripple.

#### V. CONCLUSIONS

This paper presented a methodology for obtaining voltage and current ripples in PE converters. The methodology focused on converters that have a fixed node on their output. The proposed methodology explains in detail the mathematical procedure; basically, there was used a geometric representation of the ripples for deducing the ripples. It is concluded that voltage ripple depends on current ripple and the way that ripples are related is the contribution of this paper. It is highlighted that voltage ripple is inversely proportional to the square of the switching frequency, so this kind of converters tend to have small voltage ripples. The methodology was validated through the implementation of the converter in OpenModelica software; it is concluded that the proposed methodology correctly determines the voltage and current ripples for PE converters.

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## REFERENCES

- [1] Arango, E., Ramos-Paja, C. A., Calvente, J., Giral, R., and Serna-Garcés, S. I. (2013). Asymmetrical Interleaved DC/DC Switching Converters for Photovoltaic and Fuel Cell Applications - Part 2: Control-Oriented Models. *Energies*, 6(10), 5570-5596.
- [2] Bi, Z. and Xia, W. (2010). Modeling and Simulation of Dual-Mode DC/DC Buck Converter. *Second International Conference on Computer Modeling and Simulation, Sanya, Hainan (China)*, 371-375.
- [3] Urrea-Quintero, JH., Muñoz-Galeano, N., Gómez-Echavarría, LM. (2018). Analysis and Control of Power Electronic Converters Based on a System Zeros Location Approach. In Anh Tuan, L. *Applied Modern Control* (pp.1-22). London: Intechopen.
- [4] Duong, T-D., Nguyen, M-K., Tran, T-T., Lim, Y-C., Choi, J-H. (2019). Transformerless High Step-Up DC-DC Converters with Switched-Capacitor Network. *Electronics*, 8(12), 1420.
- [5] Martínez-García, MS., de Castro, A., Sanchez, A., Garrido, J. (2019). Analysis of Resolution in Feedback Signals for Hardware-in-the-Loop Models of Power Converters. *Electronics*, 8(12), 1527.
- [6] Shaw, P. (2019). Modelling and analysis of an analogue MPPT-based PV battery charging system utilising dc-dc boost converter. *IET Renewable Power Generation*, 13(11), 1958-1967
- [7] Velilla, E., Cano, J. B., Jaramillo, F. (2019). Monitoring system to evaluate the outdoor performance of solar devices considering the power rating conditions. *Solar Energy*, 194(1), 79-85
- [8] Yang, T., Liao, Y. Discrete Sliding Mode Control Strategy for Start-Up and Steady-State of Boost Converter. (2019). *Energies*, 12(15), 2990.
- [9] Liu, J., Hu, J. and Xu, L. (2007). Dynamic Modeling and Analysis of Z Source Converter -- Derivation of AC Small Signal Model and Design-Oriented Analysis. *Power Electron. IEEE Trans.*, 22(5), 1786-1796.
- [10] Beldjajev, V. and Roasto, I. (2012). Efficiency and Voltage Characteristics of the Bi-Directional Current Doubler Rectifier. *Przeglad Elektrotechniczny*, 88(8), 124-129.
- [11] Rouzbehi, K., Miranian, A., Escaño, JM., Rakhshani, E., Shariati, N., Pouresmaeil, E. (2019). A Data-Driven Based Voltage Control Strategy for DC-DC Converters: Application to DC Microgrid. *Electronics*, 8(5), 493.
- [12] Davoudi, A., Jatskevich, J., Chapman, P. L. and Bidram, A. (2013). Multi-Resolution Modeling of Power Electronics Circuits Using Model-Order Reduction Techniques. *IEEE Transactions on Circuits and Systems I: Regular Papers*, 60(3), 810-823.
- [13] Geyer, T., Papafotiou, G., Frasca, R. and Morari, M. (2008). Constrained Optimal Control of the Step-Down DC-DC Converter. *IEEE Transactions on Power Electronics*, 23(5), 2454-2464.
- [14] Liang, T. J. and Tseng, K. C. (2005). Analysis of integrated boost-flyback step-up converter. *IEE Proceedings - Electric Power Applications*, 152(2), 217-225.
- [15] Restrepo, C., Konjedic, T., Calvente, J., Milanovic, M. and Giral, R. (2013). Fast Transitions Between Current Control Loops of the Coupled-Inductor Buck-Boost DC-DC Switching Converter. *IEEE Transactions on Power Electronics*, 28(8), 3648-3652.
- [16] Fritzson, P., Pop, A., Asghar, A., Bachmann, B., Braun, W., Braun, R., ... , Franke, R. (october, 2019). The OpenModelica Integrated Modeling, Simulation, and Optimization Environment. In *Proceedings of The American Modelica Conference 2018, Somberg Conference Center, Cambridge MA, USA*, Linköping University Electronic Press.
- [17] Dizqah, A. M., Maheri, A., Busawon, K., Fritzson, P. (2015). Standalone DC microgrids as complementarity dynamical systems: Modeling and applications. *Control Engineering Practice*, 35, 102-112.
- [18] Bartolini, A., Casella, F., Guironnet, A. (march, 2019). Towards Pan-European Power Grid Modelling in Modelica: Design Principles and a Prototype for a Reference Power System Library. In *Proceedings of the 13th International Modelica Conference, Regensburg (Germany)*, Linköping University Electronic Press.
- [19] Reid, D. (april, 2015). DQ rotating frame PI control algorithm for power inverter voltage regulation modelling and simulation using the OpenModelica platform. In *SoutheastCon 2015, Fort Lauderdale (EEUU)*, IEEE.
- [20] Murad, M. A. A., Vanfretti, L., Rokonuzzaman, M., Tuhin, R. A. (september, 2017). Enhancing engineering studies in developing countries using OpenModelica. In *2017 4th International Conference on Advances in Electrical Engineering, Dhaka (Bangladesh)*